

## Fourier domain, II

- Determining the output of an LTI system in frequency domain
- Frequency response of LTI system
- Frequency response of LTI system, example
- Frequency response of LTI system
- Frequency response of LTI system, example 2

# Determining the output of an LTI system in frequency domain

• Response of LTI system in time domain (convolution in time domain)

$$x(n) \longrightarrow h(n) \longrightarrow y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$

• Frequency domain relationship for determining the output of an LTI system

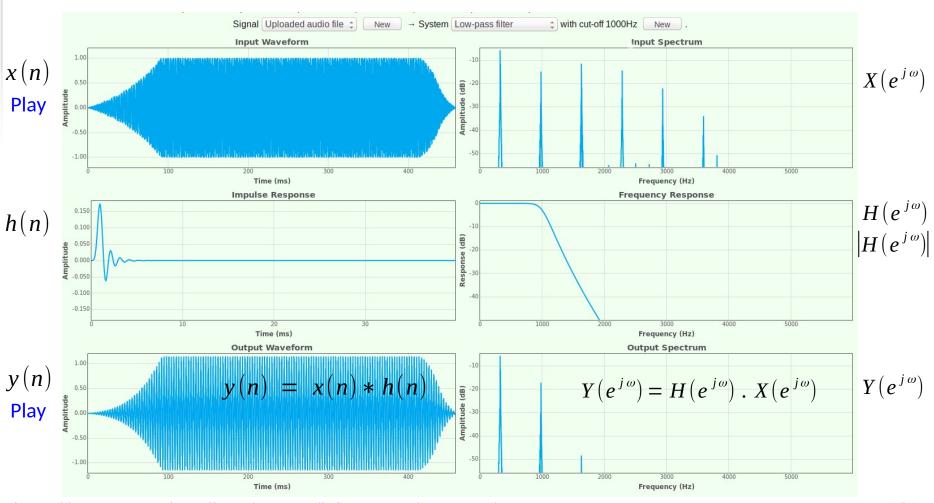
 $\rightarrow$  Convolution in time domain becomes multiplication in frequency domain

$$X(e^{j\omega}) \longrightarrow H(e^{j\omega}) \longrightarrow Y(e^{j\omega})$$

$$X(e^{j\omega}) = DTFT[x(n)]$$
$$H(e^{j\omega}) = DTFT[h(n)]$$
$$Y(e^{j\omega}) = DTFT[y(n)]$$

Spectrum of the output signal  $Y(e^{j\omega}) = H(e^{j\omega}) \cdot X(e^{j\omega})$ 

Output signal  $y(n) = IDTFT[Y(e^{j\omega})]$ 



<u>http://www.speechandhearing.net/laboratory/esystem/</u>

Course: 63744

• Impulse response h(n) completely describes LTI system

$$x(n) \longrightarrow h(n) \longrightarrow y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k)$$

• Complex sinusoid as input signal

$$x(n) = e^{j\omega n} = \cos(\omega n) + j\sin(\omega n)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)e^{j\omega(n-k)}$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)e^{-j\omega k} \cdot e^{j\omega n}$$

$$y(n) = H(e^{j\omega}) \cdot e^{j\omega n} \rightarrow$$

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h(k) e^{-j\omega k}$$

Frequency response of LTI system is the DTFT of the impulse response h(n) of the system

ightarrow Output is sinusoid scaled by DTFT of h(n) at  $\omega$ 

• If the input signal is complex sinusoid  $x(n) = e^{j\omega n}$   $e^{j\omega n} \longrightarrow H(e^{j\omega}) \longrightarrow y(n) = H(e^{j\omega}) \cdot e^{j\omega n}$ • Since frequency response is  $H(e^{j\omega}) = H_R(e^{j\omega}) + jH_I(e^{j\omega})$   $H(e^{j\omega}) = |H(e^{j\omega})| \cdot e^{j\theta(\omega)}$  (polar notation !) where  $|H(e^{j\omega})|$  is amplitude response (gain) of the system  $\theta(\omega)$  is phase response (phase shift) of the system

follows:  $y(n) = |H(e^{j\omega})| \cdot e^{j(\omega n + \theta(\omega))}$ 

The output of the system with impulse response h(n) is the same sinusoidscaled by $|H(e^{j\omega})|$ and phase-shifted by $\theta(\omega)$ where $H(e^{j\omega}) = \text{DTFT}[h(n)]$ 

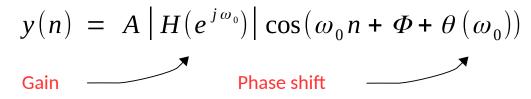
- In practice signals, x(n), h(n), are real
- Example

$$x(n) = A\cos(\omega_0 n + \Phi)$$

(Derivation)

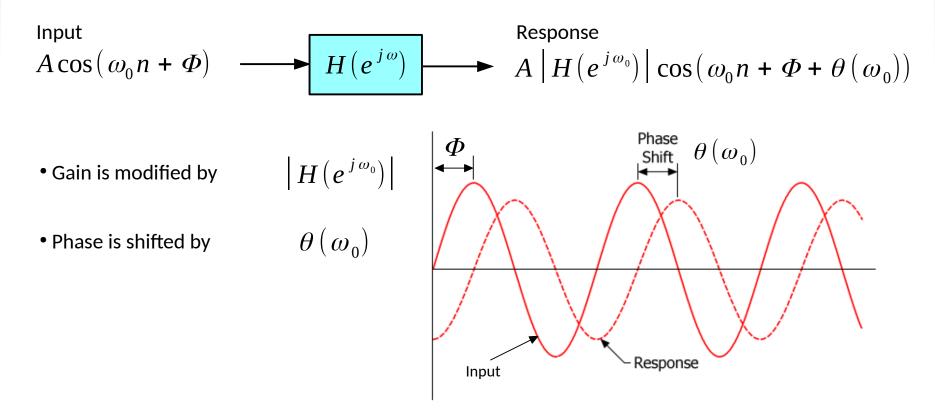
$$\begin{aligned} x(n) &= \frac{A}{2} \left( e^{j(\omega_0 n + \Phi)} + e^{-j(\omega_0 n + \Phi)} \right) \\ x(n) &= \frac{A}{2} e^{j\Phi} e^{j\omega_0 n} + \frac{A}{2} e^{-j\Phi} e^{-j\omega_0 n} \\ y(n) &= \frac{A}{2} e^{j\Phi} H(e^{j\omega_0}) e^{j\omega_0 n} + \frac{A}{2} e^{-j\Phi} H(e^{-j\omega_0}) e^{-j\omega_0 n} \\ \text{For real } h(n) \Rightarrow H(e^{-j\omega}) = H^*(e^{j\omega}) = \left| H(e^{j\omega}) \right| e^{-j\Theta(\omega)} \end{aligned}$$

Output:



Course: 63744

• Passing a real sinusoid through an LTI system with real h(n)



• LTI system

$$x(n) = e^{j\omega n} \longrightarrow H(e^{j\omega}) \longrightarrow y(n) = H(e^{j\omega}) \cdot e^{j\omega n}$$

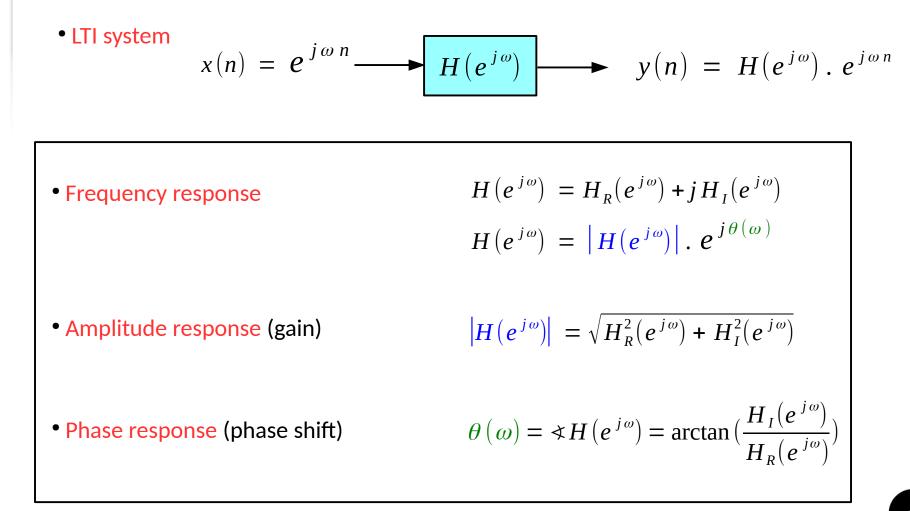
• Arbitrary input signal as a sum of sinusoids

$$x(n) = \sum_{i=1}^{L} A_i \cos(\omega_i n + \Phi_i)$$

• The output signal is the sum of responses

$$y(n) = \sum_{i=1}^{L} A_i |H(e^{j\omega_i})| \cos(\omega_i n + \Phi_i + \theta(\omega_i))$$

Influence of the system to the *i*-th component of the signal



• Derive frequency response, amplitude response, and phase response of the following moving average filter

$$y(n) = \frac{1}{M} \sum_{l=0}^{M-1} x(n-l), \quad M = 5$$

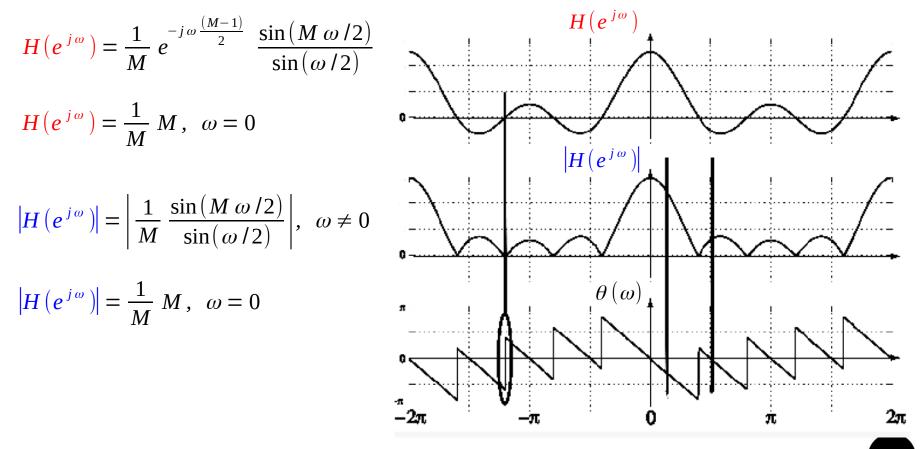
$$h(n) = \{1/5, 1/5, 1/5, 1/5, 1/5\} \qquad \qquad \begin{array}{c} h(n) \\ 1/M \\ 0 \\ 0 \\ -1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 6n \end{array}$$

• Study the output y(n) for the following cases:  $\omega_0 = 0.1 \pi$   $\omega_1 = 0.5 \pi$ 

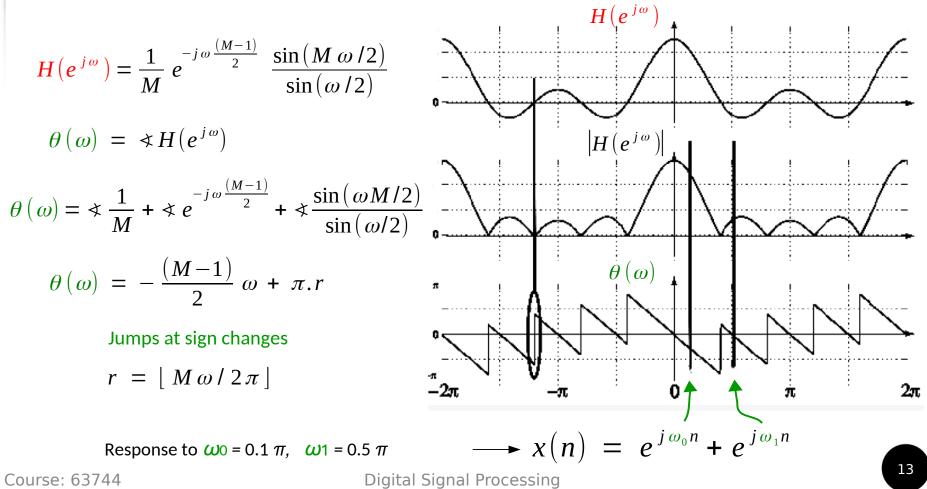
• Frequency response of moving average filter

$$H(e^{j\omega}) = \text{DTFT}[h(n)] = \sum_{n=-\infty}^{\infty} h(n)e^{-j\omega n} = \frac{1}{M} \sum_{n=0}^{M-1} e^{-j\omega n}$$
$$H(e^{j\omega}) = \frac{1}{M} (1 + e^{-j\omega} + e^{-j\omega 2} + \dots + e^{-j\omega (M-1)}) = \frac{1}{M} \frac{1 - e^{-j\omega M}}{1 - e^{-j\omega}}$$
$$H(e^{j\omega}) = \frac{1}{M} e^{-j\omega \frac{(M-1)}{2}} \frac{\sin(M\omega/2)}{\sin(\omega/2)}, \quad \omega \neq 0$$
$$H(e^{j\omega}) = \frac{1}{M} M, \quad \omega = 0$$

• Frequency response and amplitude response, *M* = 5



• Frequency response and phase response



- Moving average, M = 5
- Input

 $x(n) = e^{j\omega_0 n} + e^{j\omega_1 n}$ 

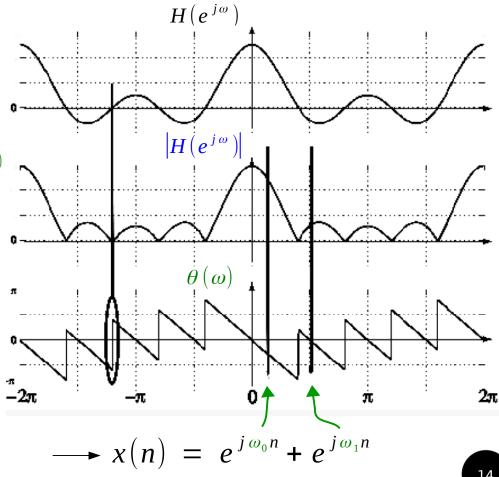
$$\omega_0 = 0.1 \pi \rightarrow H(e^{j\omega_0}) \approx 0.8 e^{j\theta(\omega_0)}$$
$$\omega_1 = 0.5 \pi \rightarrow H(e^{j\omega_1}) \approx (-) 0.2 e^{j\theta(\omega_1)}$$

$$\theta(\omega_0) = -2 \omega_0 = -2(0.1 \pi) = -0.2 \pi$$
  
 $\theta(\omega_1) = -2 \omega_1 = -2(0.5 \pi) = -\pi$ 

• Output

$$y(n) = H(e^{j\omega_0}) e^{j\omega_0 n} + H(e^{j\omega_1}) e^{j\omega_1 n}$$

Response to  $\omega_0 = 0.1 \pi$ ,  $\omega_1 = 0.5 \pi$ 



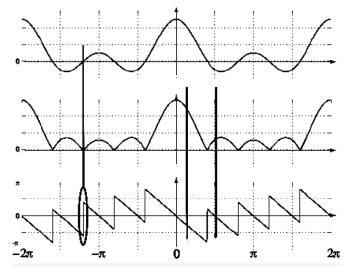
Course: 63744

Frequency response of LTI system, example

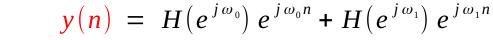
- Moving average, M = 5
- Input

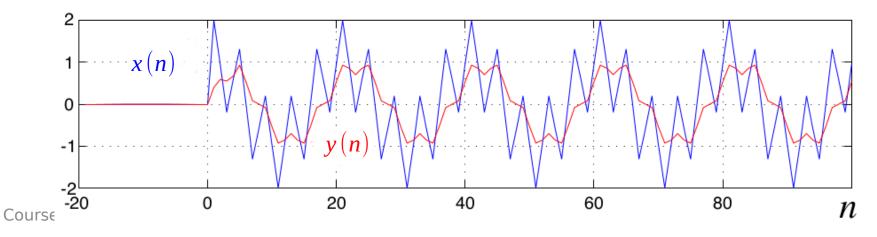
$$x(n) = e^{j\omega_0 n} + e^{j\omega_1 n}$$

$$\omega_0 = 0.1 \pi \rightarrow H(e^{j\omega_0}) \approx 0.8 e^{j\theta(\omega_0)}$$
$$\omega_1 = 0.5 \pi \rightarrow H(e^{j\omega_1}) \approx (-) 0.2 e^{j\theta(\omega_1)}$$



• Output





#### • Example of exam task

Derive frequency, amplitude and phase response of the following LTI system. Also sketch the amplitude and phase response.

y(n) = x(n) - x(n-1)

(Hint: Derive the impulse response h(n) of the system first)

#### • Example of exam task

Derive frequency, amplitude and phase response of the following LTI system. Also sketch the amplitude and phase response.

$$y(n) = x(n) + x(n-1)$$

(Hint: Derive the impulse response h(n) of the system first)



## Frequency response of LTI system, example 2, laboratory

• Derive frequency response, amplitude response, and phase response of an LTI system which is defined by the following impulse response

- Draw amplitude and phase response when *a* = 0.6
- Calculate values of the output signal, y(n), amplitude of the output signal,  $|H(e^{j\omega})|$ , and phase of the output signal,  $\theta(\omega)$ , in the case of following input signal

$$x(n) = 5 + 12\sin(\frac{\pi}{2}n) - 20\cos(\pi n + \frac{\pi}{4})$$

• Impulse response and frequency response of the system

$$h(n) = a^{n}u(n), \quad a = 0.6$$
  
 $h(n)$   
 $a = 0.6$   
 $h(n)$   
 $a = 0.6$ 

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h(n) e^{-j\omega n} = \sum_{n=-\infty}^{\infty} a^n u(n) e^{-j\omega n}$$
$$H(e^{j\omega}) = \sum_{n=0}^{\infty} (a e^{-j\omega})^n = \frac{1}{1-a e^{-j\omega}}$$

• Frequency, amplitude, and phase response of the system

$$h(n) = a^{n}u(n), \quad a = 0.6$$

$$\int_{n=-\infty}^{\infty} e^{n}u(n) e^{-j\omega n} = \sum_{n=0}^{\infty} (ae^{-j\omega})^{n} = \frac{1}{1-ae^{-j\omega}}$$

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} a^{n}u(n) e^{-j\omega n} = \sum_{n=0}^{\infty} (ae^{-j\omega})^{n} = \frac{1}{1-ae^{-j\omega}}$$

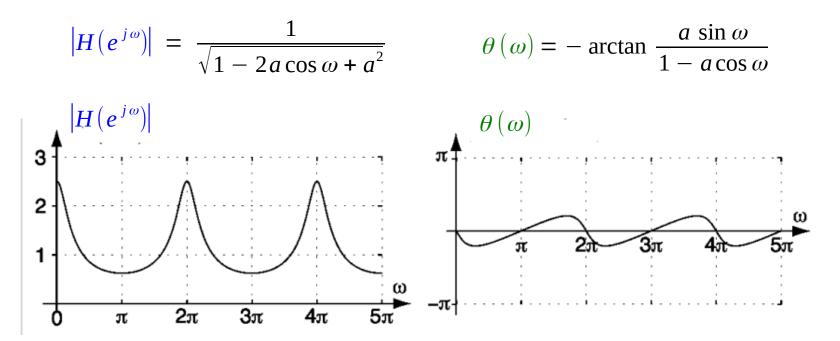
$$H(e^{j\omega}) = X_{R}(e^{j\omega}) + X_{I}(e^{j\omega}) = \frac{1-a\cos\omega - ja\sin\omega}{1-2a\cos\omega + a^{2}}$$

$$|H(e^{j\omega})| = \frac{1}{\sqrt{(1-a\cos\omega)^{2} + (a\sin\omega)^{2}}} = \frac{1}{\sqrt{1-2a\cos\omega + a^{2}}}$$

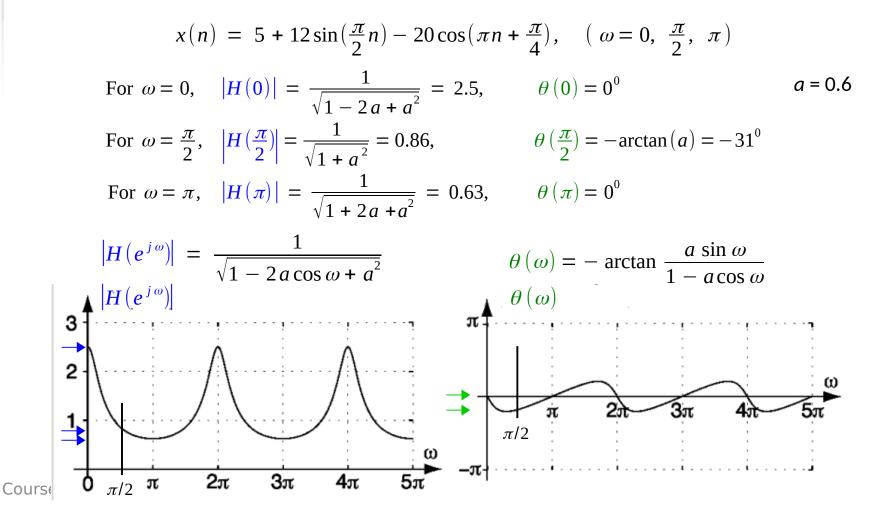
$$\theta(\omega) = -\arctan\frac{a\sin\omega}{1-a\cos\omega}$$

#### • Draw the amplitude and phase response of the system

$$h(n) = a^n u(n), \qquad a = 0.6$$



• Determine the output y(n) of the system to the following input signal



• Determine the output y(n) of the system to the following input signal

$$x(n) = 5 + 12\sin(\frac{\pi}{2}n) - 20\cos(\pi n + \frac{\pi}{4}), \quad (\omega = 0, \frac{\pi}{2}, \pi)$$
  
For  $\omega = 0$ ,  $|H(0)| = \frac{1}{\sqrt{1 - 2a + a^2}} = 2.5$ ,  $\theta(0) = 0^0$   $a = 0.6$   
For  $\omega = \frac{\pi}{2}$ ,  $|H(\frac{\pi}{2})| = \frac{1}{\sqrt{1 + a^2}} = 0.86$ ,  $\theta(\frac{\pi}{2}) = -\arctan(a) = -31^0$   
For  $\omega = \pi$ ,  $|H(\pi)| = \frac{1}{\sqrt{1 + 2a + a^2}} = 0.63$ ,  $\theta(\pi) = 0^0$ 

$$y(n) = 5 \left| H(0) \right| + 12 \left| H(\frac{\pi}{2}) \right| \sin\left[\frac{\pi}{2}n + \theta\left(\frac{\pi}{2}\right)\right] - 20 \left| H(\pi) \right| \cos\left[\pi n + \frac{\pi}{4} + \theta\left(\pi\right)\right]$$